OPTICS LAB TUTORIAL: Oscilloscope and Spectrum Analyzer

M.P. Hasselbeck

Oscilloscope or Multimeter?





Multimeter

- Battery powered
- Hand-held
- Very portable
- Variety of measurements possible

DC Voltage Continuity

AC Voltage Resistance

DC Current Capacitance

AC Current



Multimeter

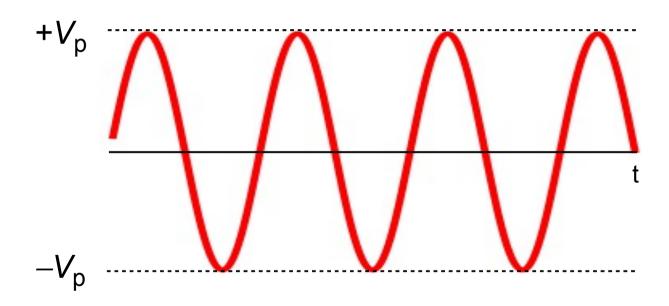
Measurement given as a single number

What about signals that change as a function of time?



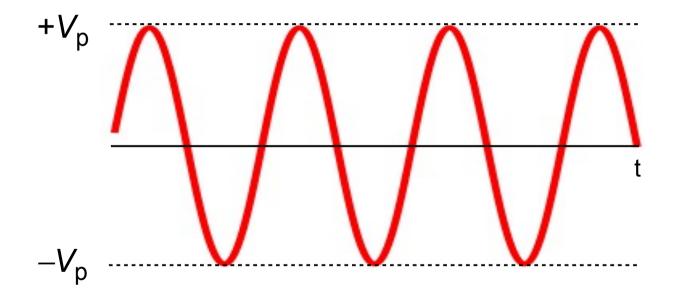
Periodic, time-varying signals can sometimes be characterized by a single number: Root-Mean-Square (RMS)

$$V(t) = V_p \sin(\omega t)$$

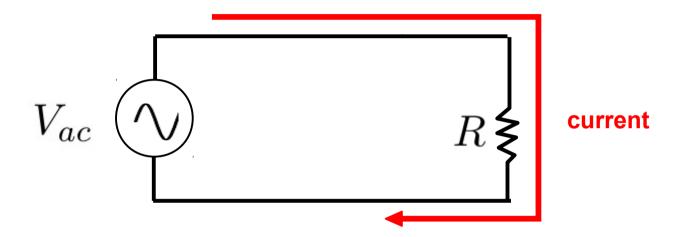


The average voltage of a pure sine wave is identically zero.

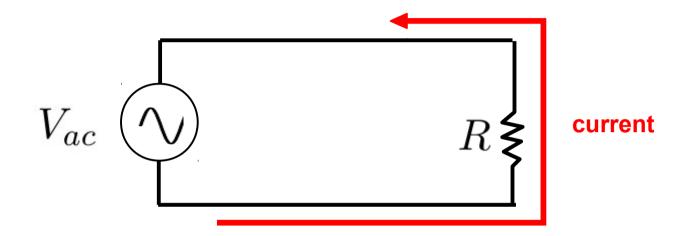
$$V(t) = V_p \sin(\omega t)$$



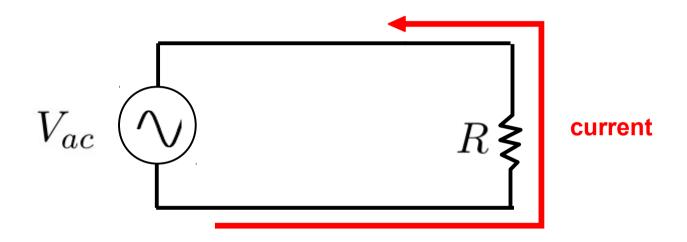
We know that an AC voltage can deliver plenty of power to a load



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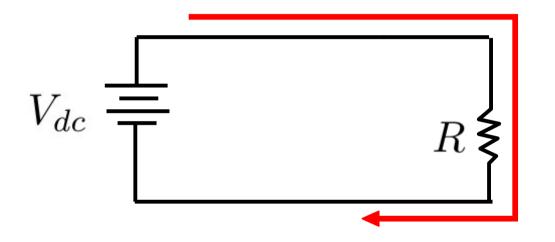


We know that an AC voltage can deliver plenty of power to a load



How do we calculate this power if the average voltage and current is zero?

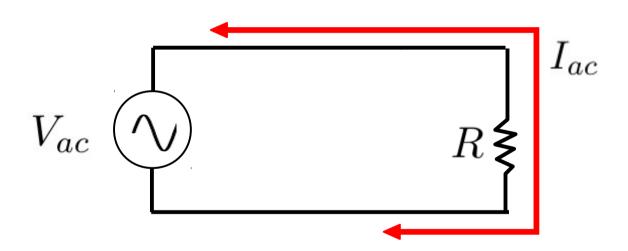
This is easy in a DC circuit: Use Ohm's Law



POWER =
$$\frac{V_{dc}^2}{R}$$

Power dissipated in an AC circuit is time dependent:

$$P(t) = V_{ac}I_{ac} = \frac{V_{ac}^2}{R}$$



What is the energy delivered in **one period**? Temporally integrate the power over one period:

$$\frac{1}{T} \int_0^T P(t)dt = \frac{1}{TR} \int_0^T V_p^2 \sin(\omega t)^2 dt = \frac{V_p^2}{2R}$$

DC power dissipation =
$$\frac{V_{dc}^2}{R}$$

AC power dissipation =
$$\frac{V_p^2}{2R}$$

AC voltage producing power dissipation equivalent $V_{RMS} = \frac{V_p}{\sqrt{2}}$

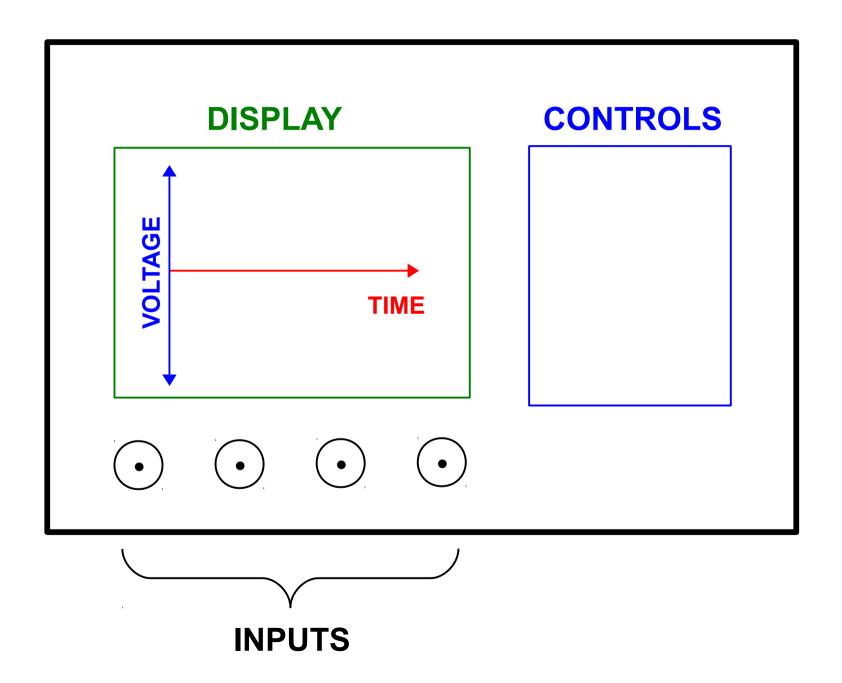
$$V_{RMS} = \frac{V_p}{\sqrt{2}}$$

- An RMS measurement assumes a stable, periodic signal
- Characterized by a single value of voltage, current
- Measured with a multimeter or oscilloscope

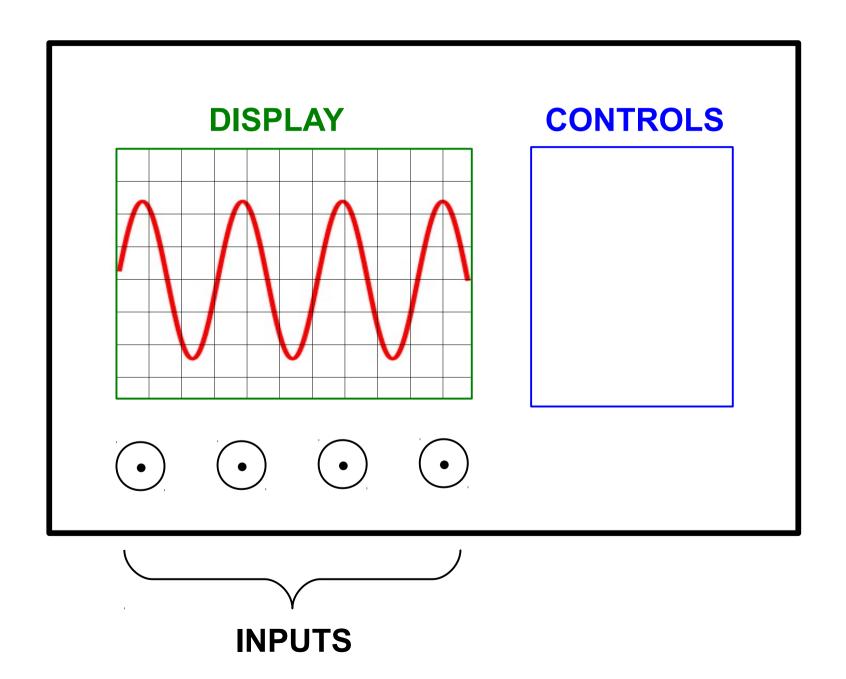
The situation is often not that convenient!



OSCILLOSCOPE



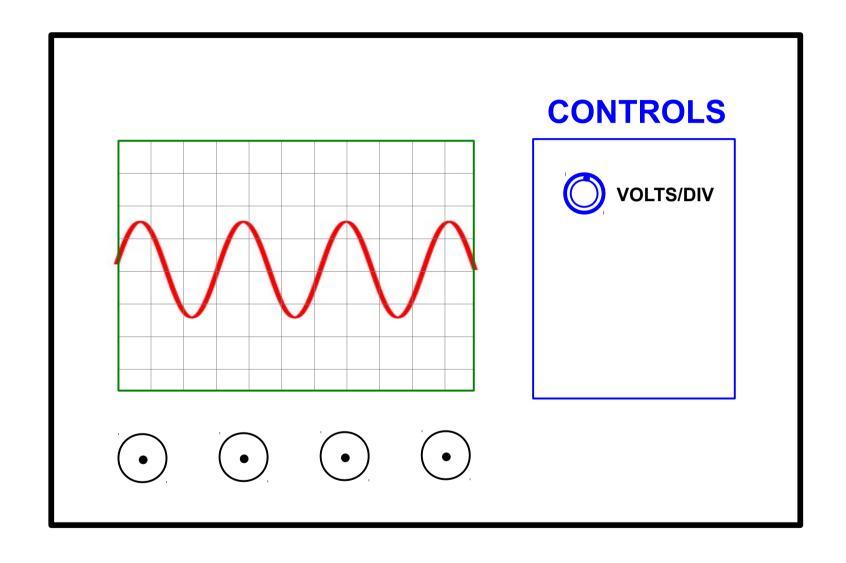
OSCILLOSCOPE

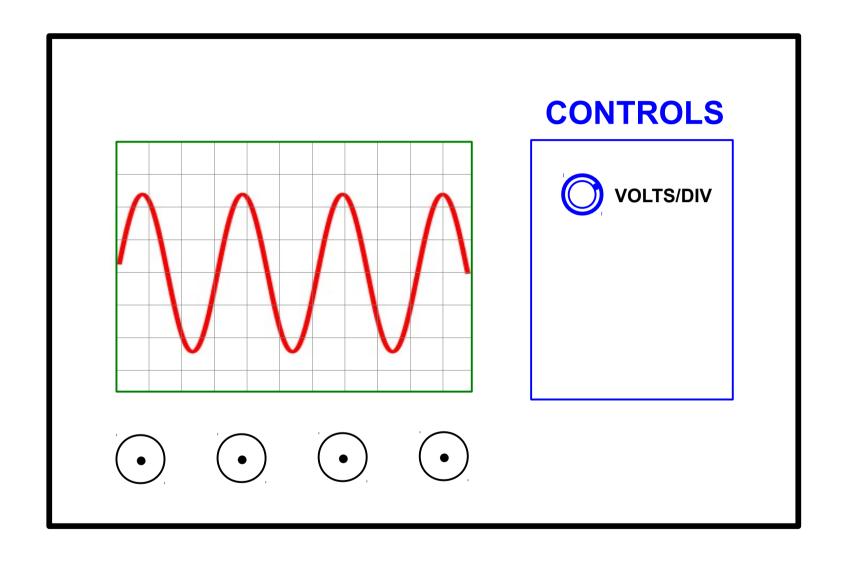


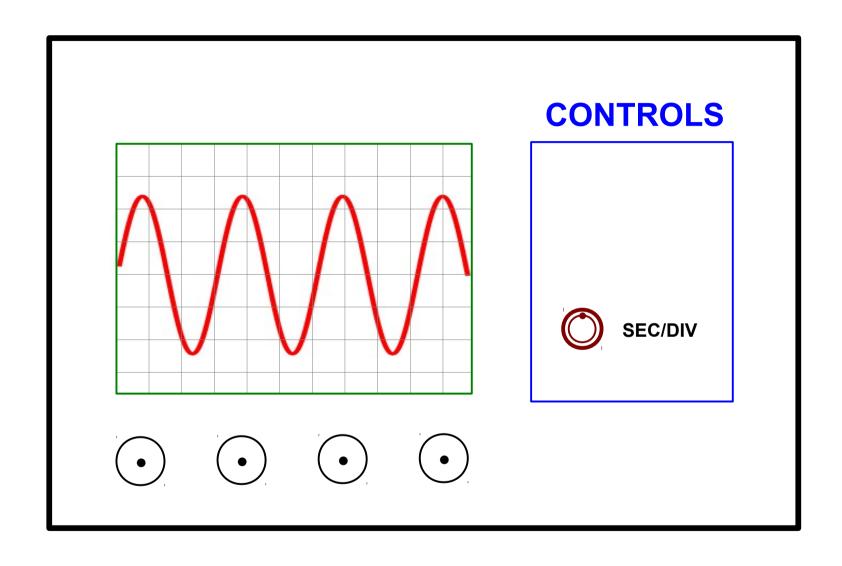
ANALOG: Cathode ray tube, swept electron beam

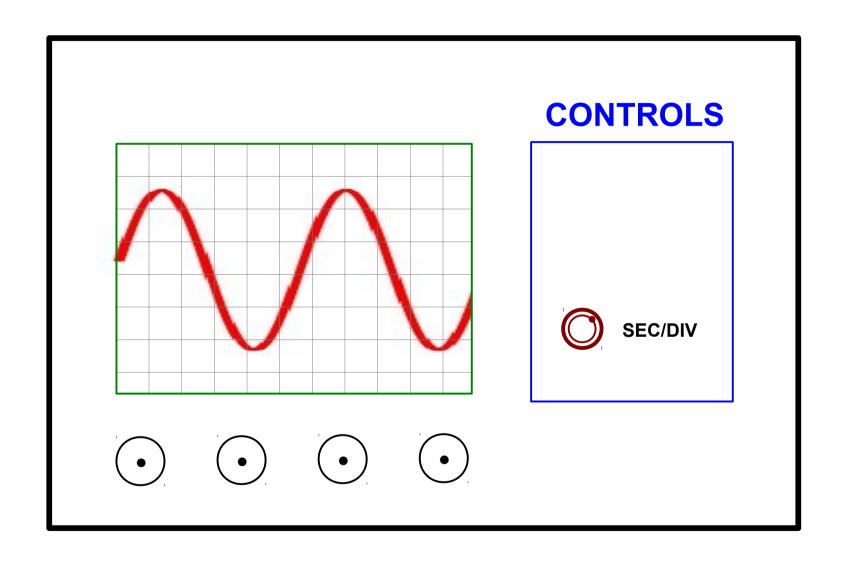
DIGITAL: A/D converter, LCD display

Although physical operation is completely different, controls are nearly identical

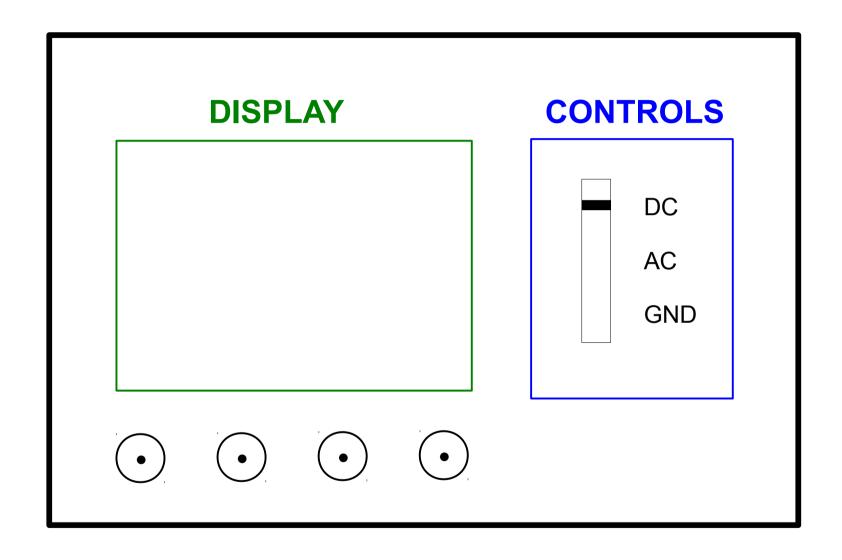




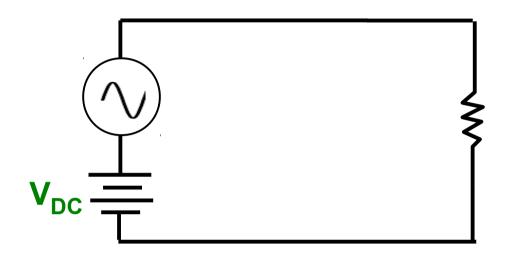


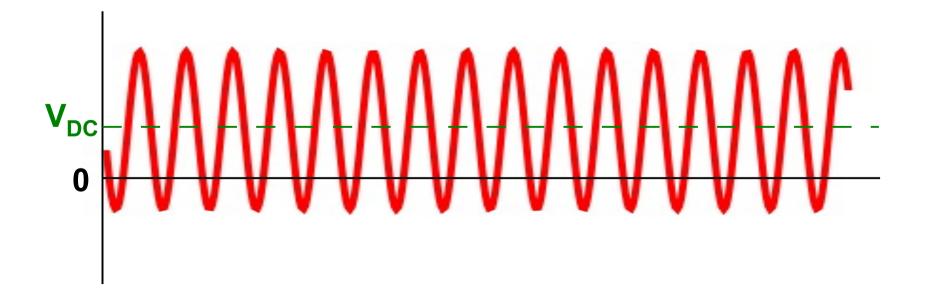


DC coupling, AC coupling, and Ground

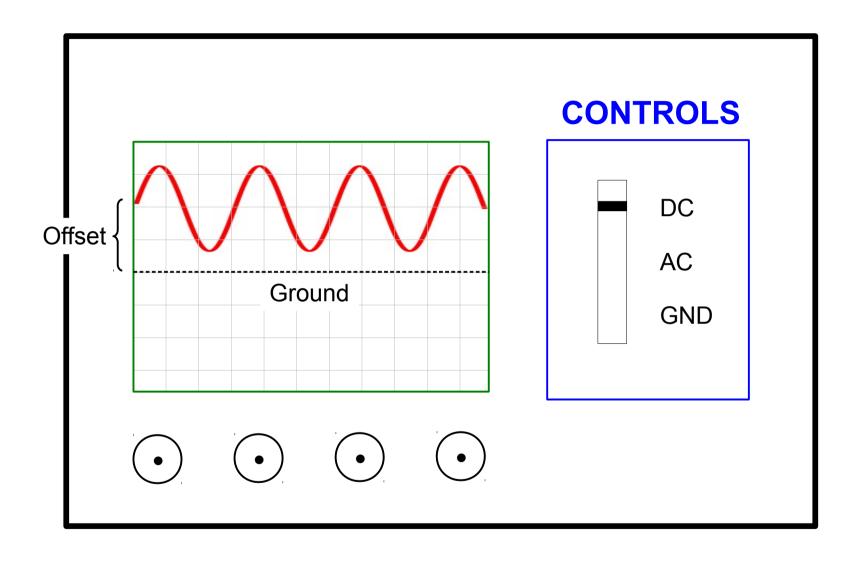


EXAMPLE: Sinusoidal wave source + DC offset

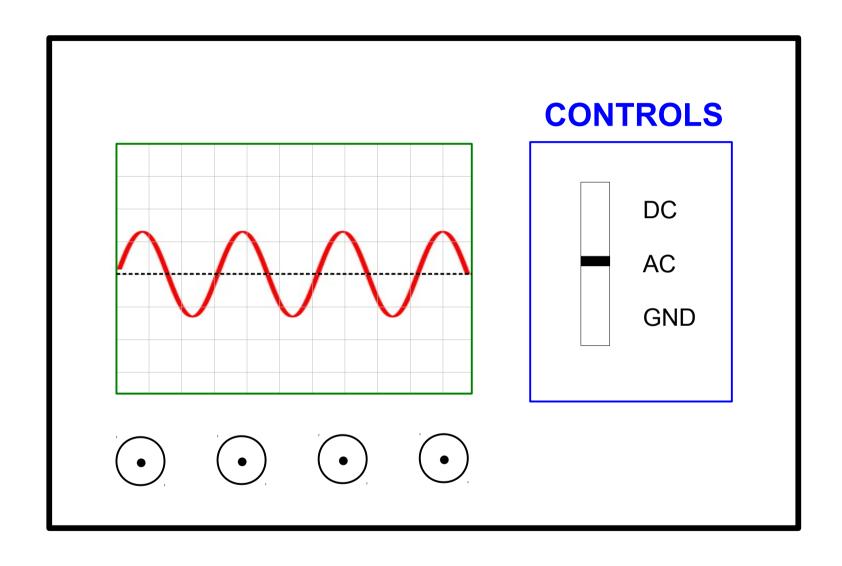




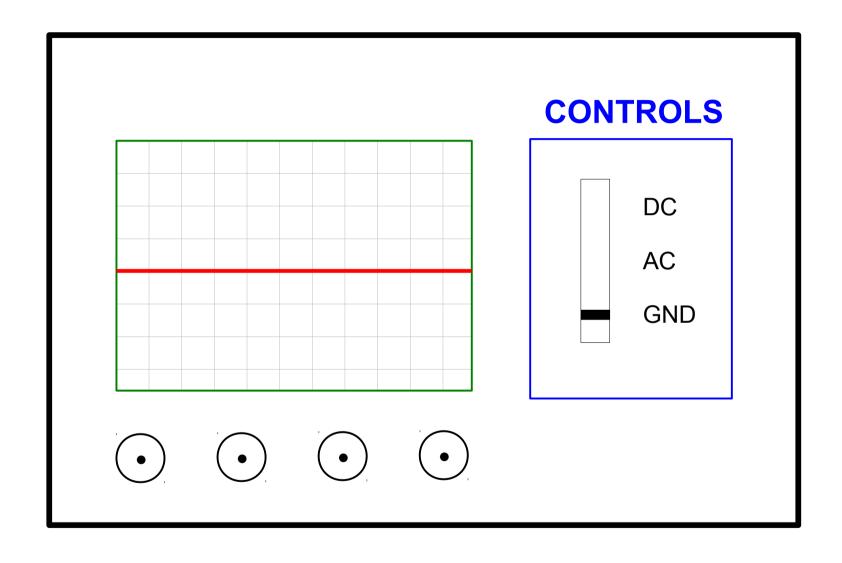
DC COUPLING



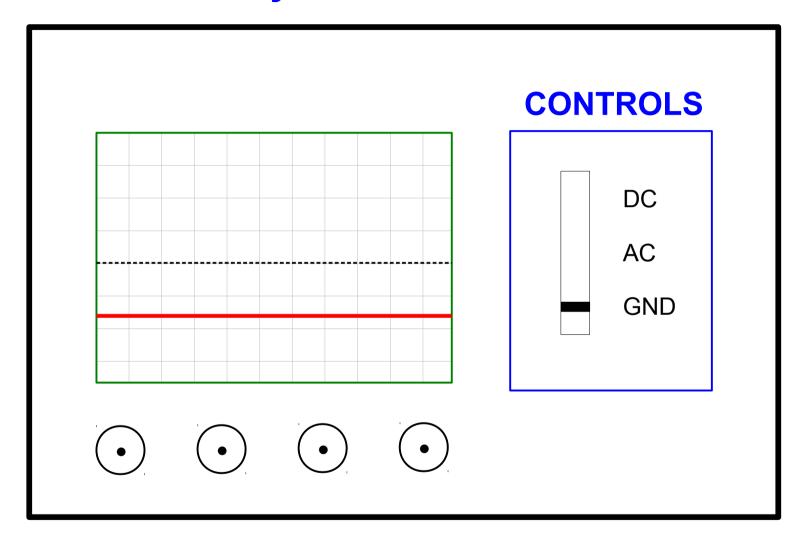
AC COUPLING



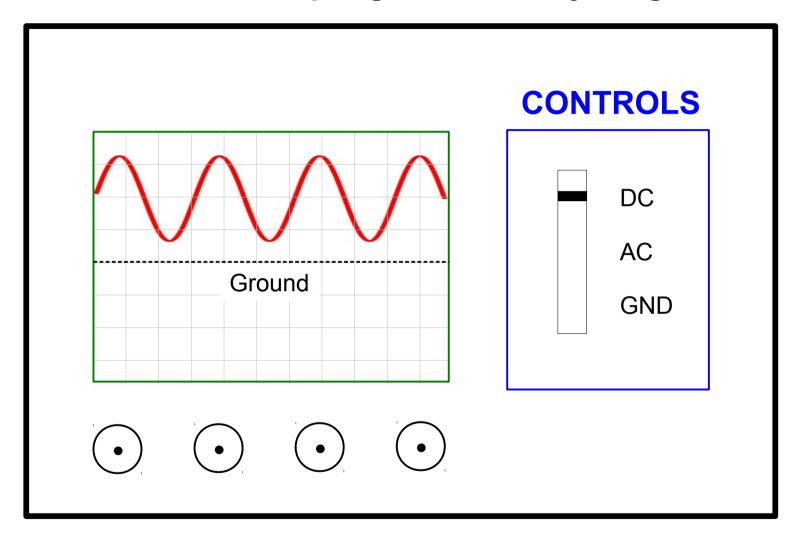
GROUND: Defines location of 0 Volts



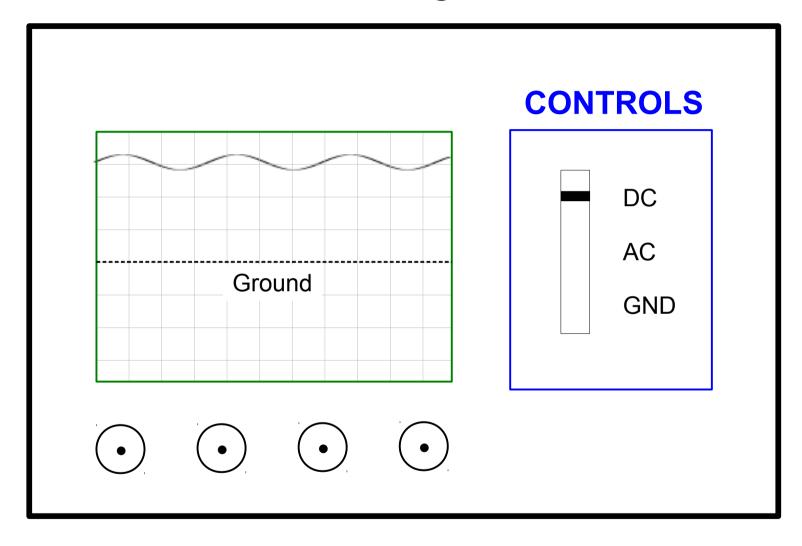
GROUND can be positioned at any convenient level



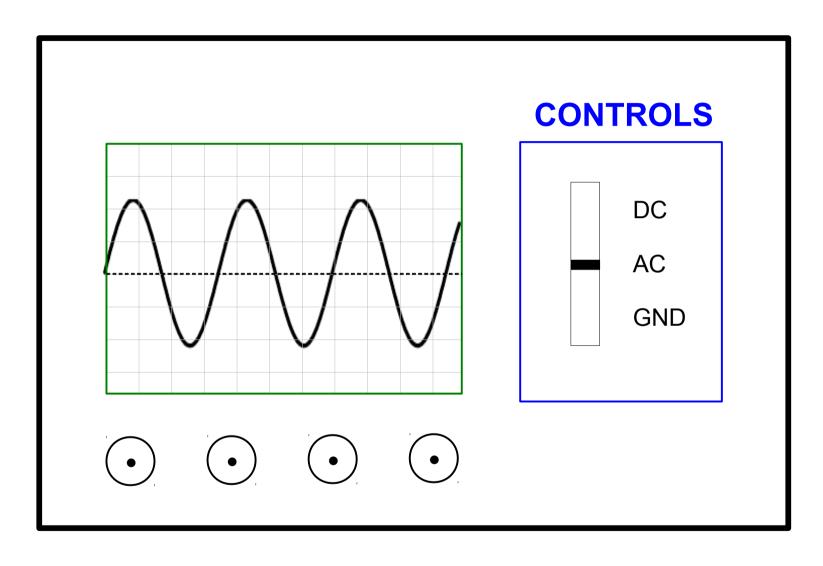
Why bother with AC coupling when DC coupling shows everything?



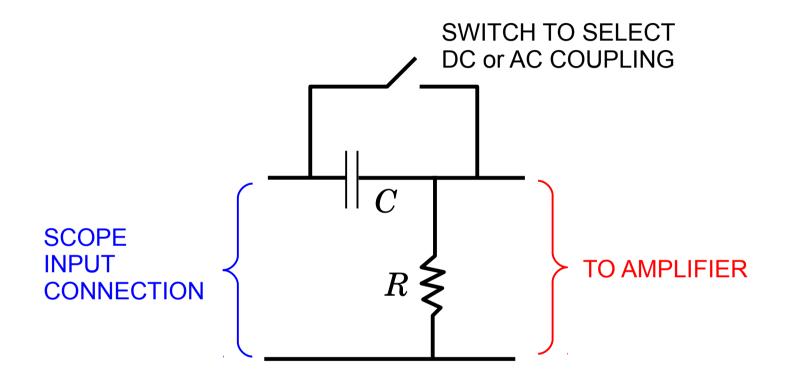
Often we have very weak modulation of a DC signal



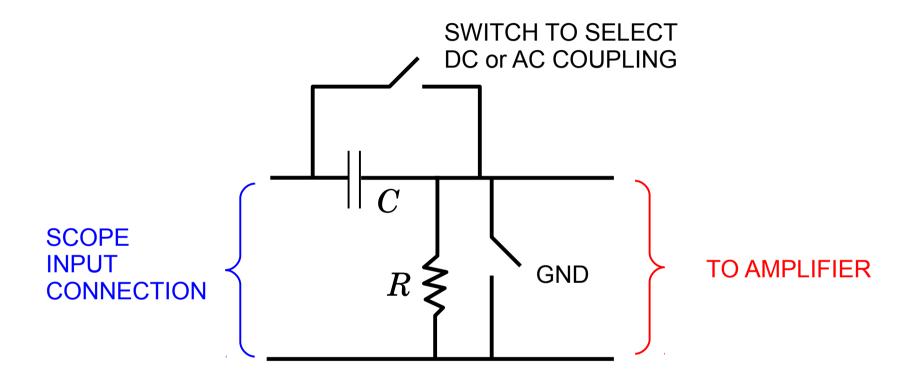
AC couple and change the vertical scale



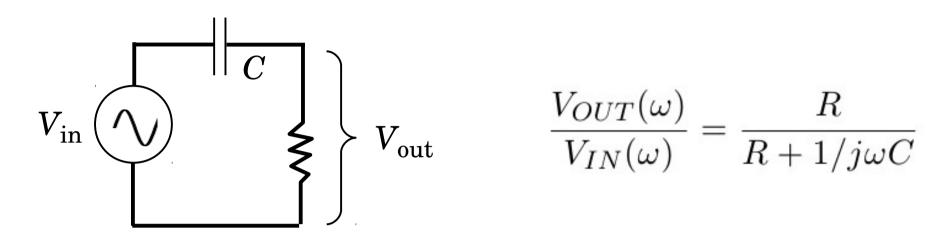
AC coupling implemented with an RC high-pass filter



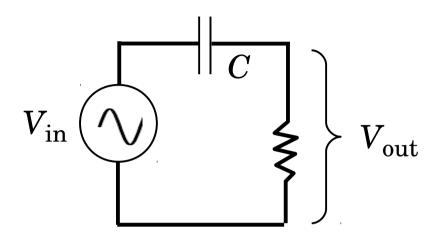
AC coupling implemented with an RC high-pass filter



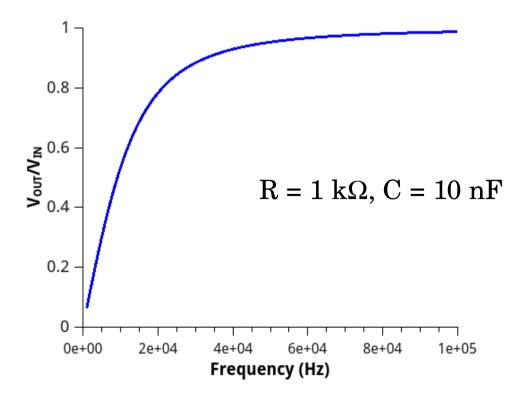
Harmonic analysis of RC high-pass filter

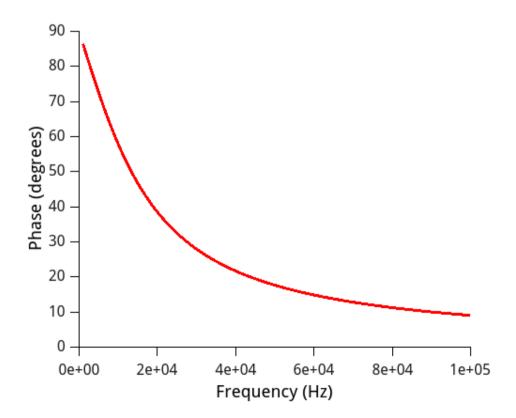


Harmonic analysis of RC high-pass filter

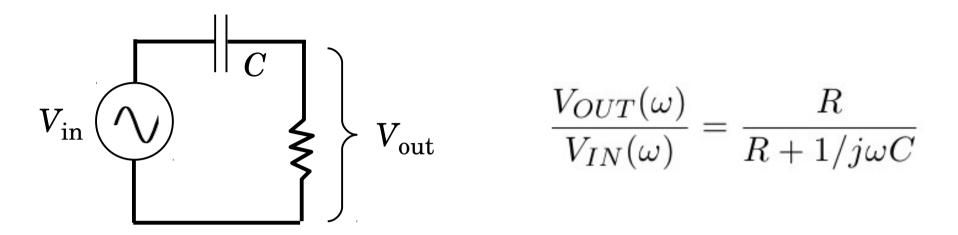


$$\frac{V_{OUT}(\omega)}{V_{IN}(\omega)} = \frac{R}{R + 1/j\omega C}$$





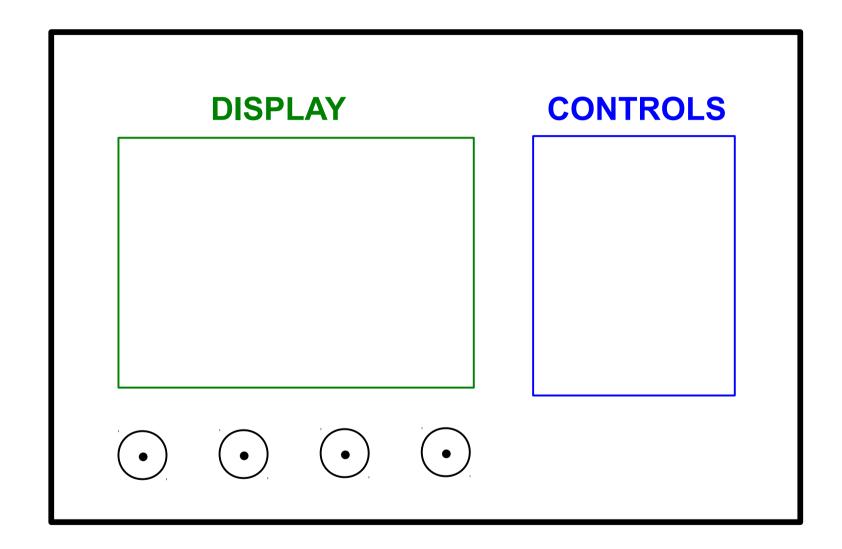
Harmonic analysis of RC high-pass filter



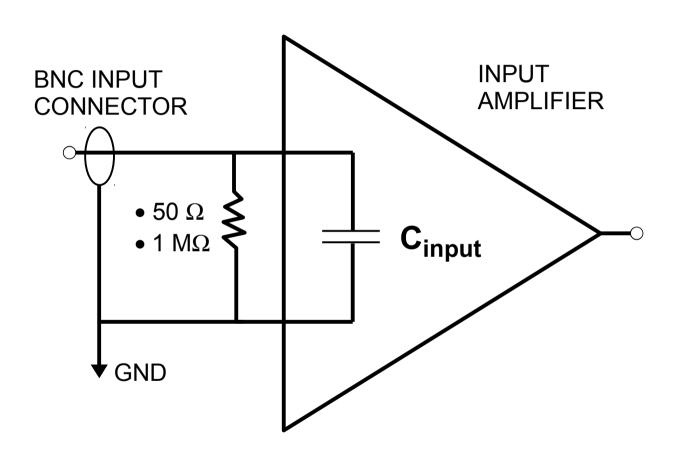
A typical oscilloscope has an *RC* high-pass cutoff in the range 1—10 Hz when AC coupling is used

Be careful when measuring slow signals: AC coupling blocks more than just DC

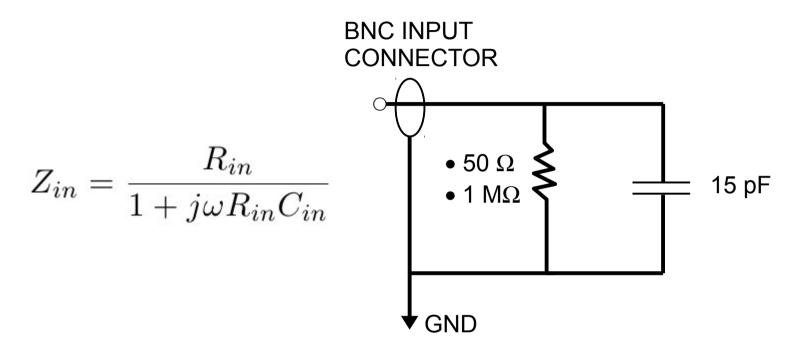
INPUT RESISTANCE: 50 Ω or 1 M Ω ?



All oscilloscopes have stray (unavoidable) capacitance at the input terminals: $C_{input} = 15-20 \text{ pF}$



All oscilloscopes have stray (unavoidable) capacitance at the input terminals

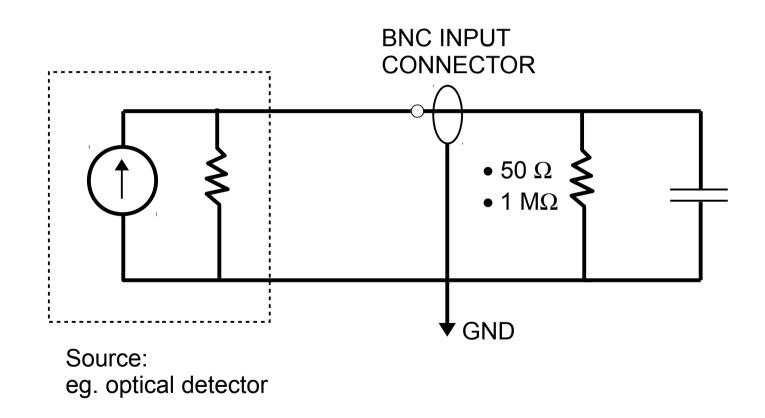


- 1 MΩ rolloff ~ 8 kHz
- 50 Ω rolloff ~ 160 MHz

Compensation possible with scope probe

Why do we use 1 $M\Omega$ if frequency response is so low?

ANSWER: Signal level (voltage) will drop enormously at 50 Ω unless source can provide enough current



TRIGGERING

Auto: Scope gives continually updated display

Normal: User controls when the slope triggers; Level, Slope

Trigger source: Channel 1, Channel 2, etc

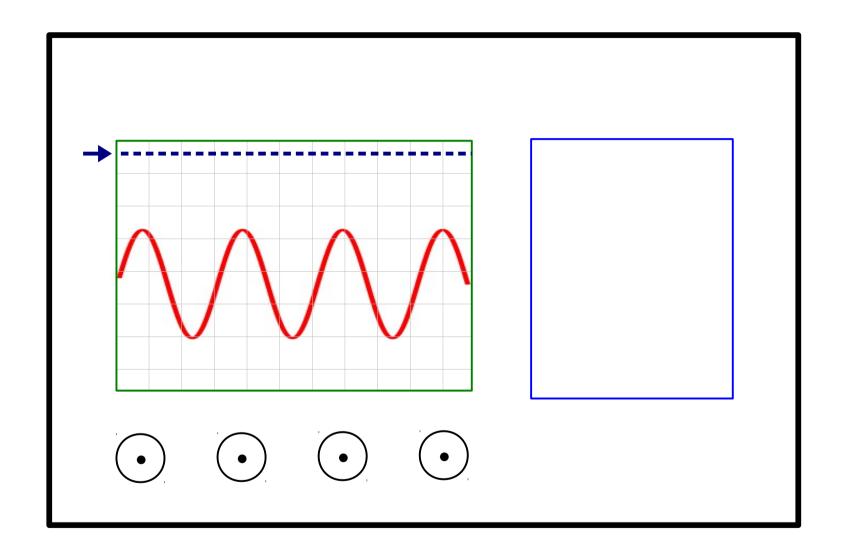
Line: Triggers on 60 Hz AC

Single event

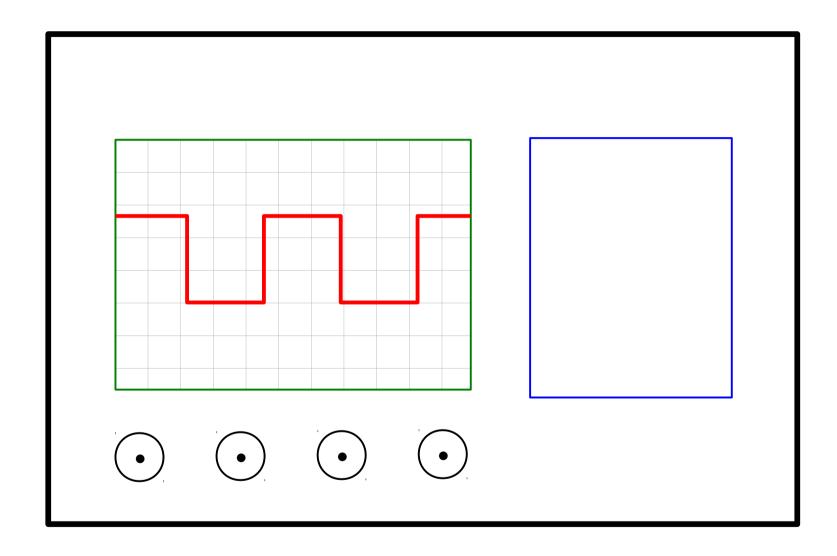
External

Use Auto-Set only when all else fails!

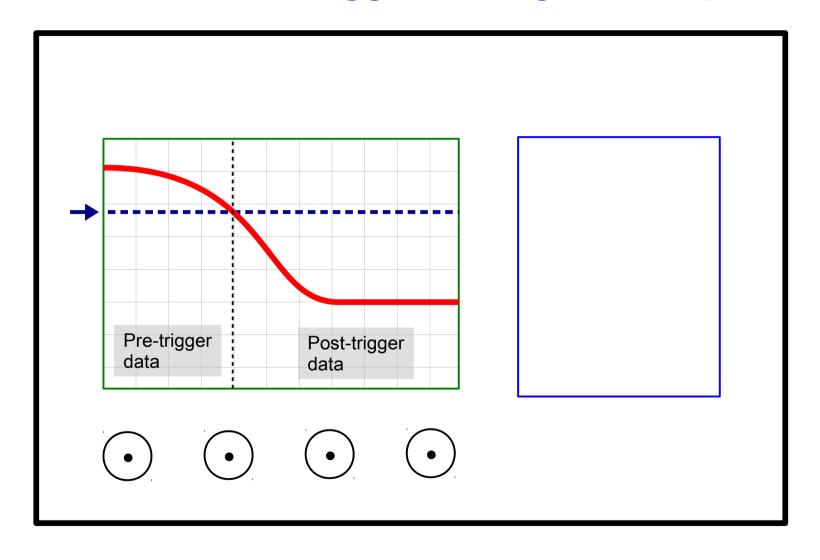
Setting normal trigger level



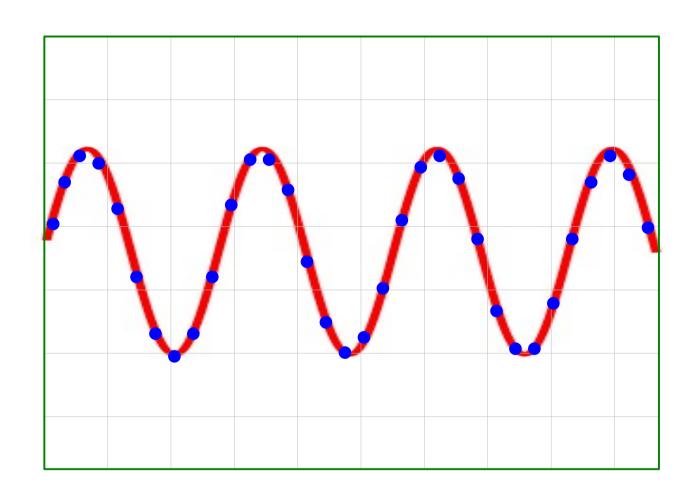
Example: Measure fall time of square wave

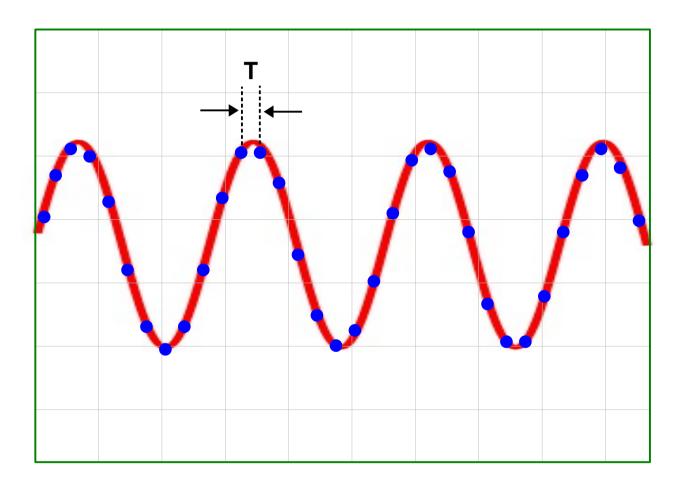


SOLUTION: Trigger on negative slope



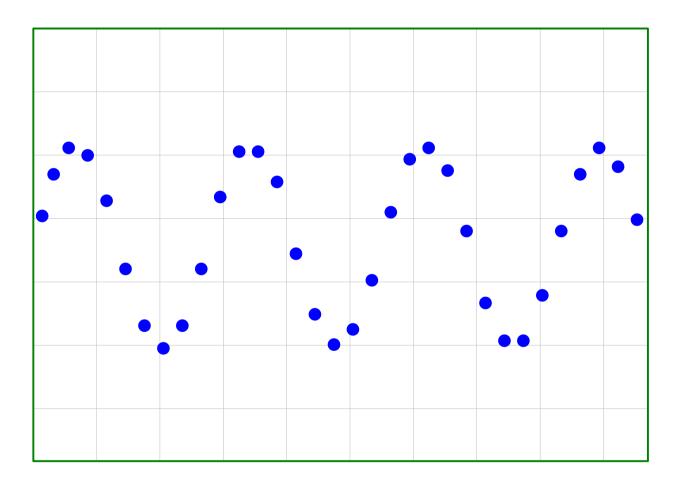
DIGITAL SCOPE: SAMPLING BANDWIDTH





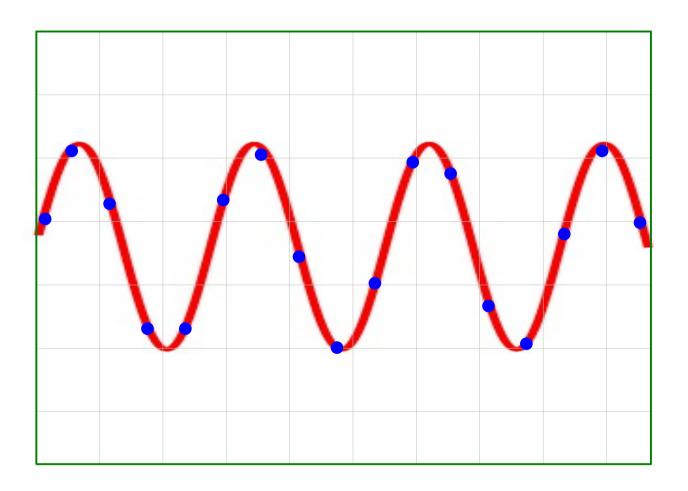
Sample spacing: **T** (sec)

Sampling bandwidth = 1 / **T** (samples/sec)

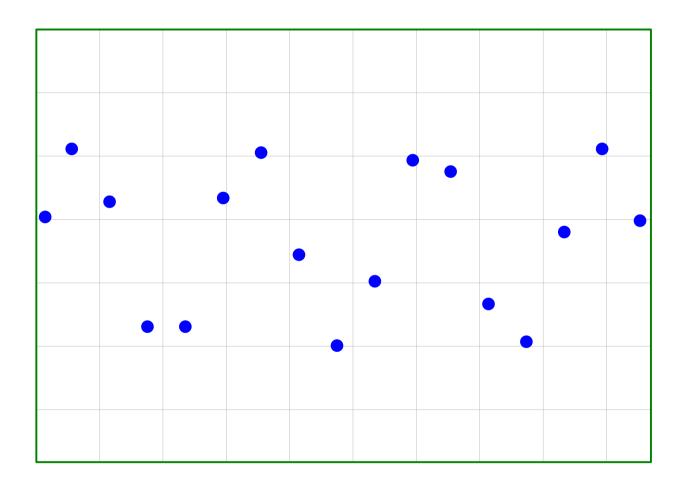


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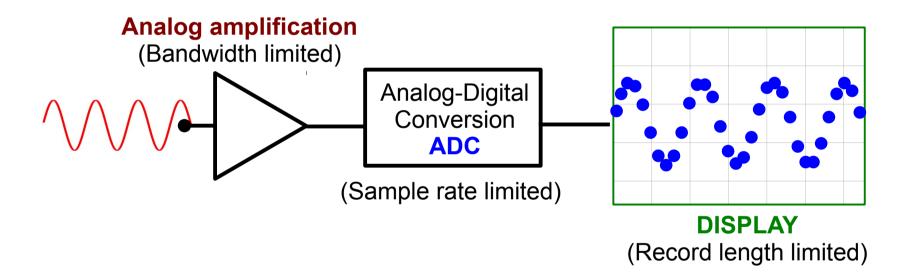


Reduce sample bandwidth $2x \implies$ Increase period 2x

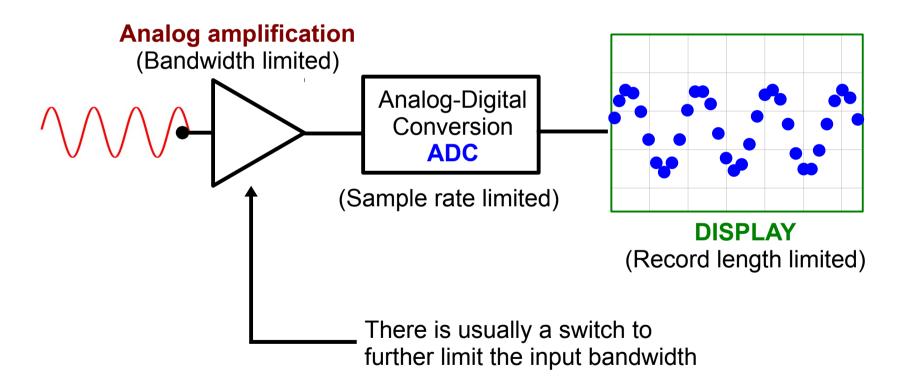


Reduce sample bandwidth $2x \implies$ Increase period 2x

ANALOG BANDWIDTH \neq SAMPLING BANDWIDTH



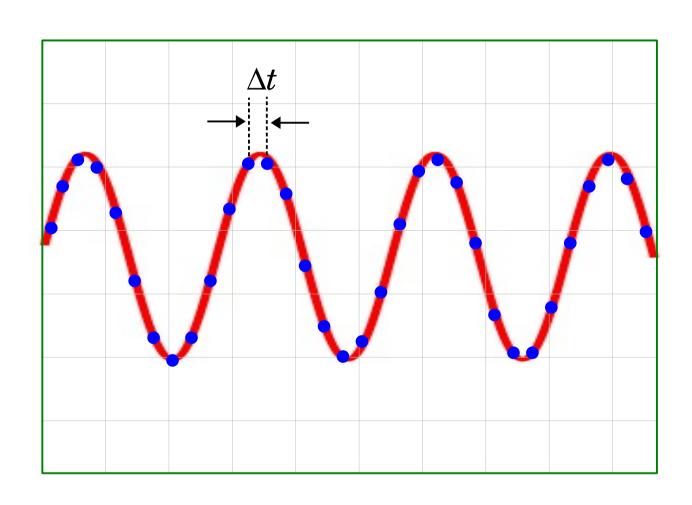
ANALOG BANDWIDTH \neq SAMPLING BANDWIDTH



Nyquist theorem Sampling theorem

Temporal spacing of signal sampling

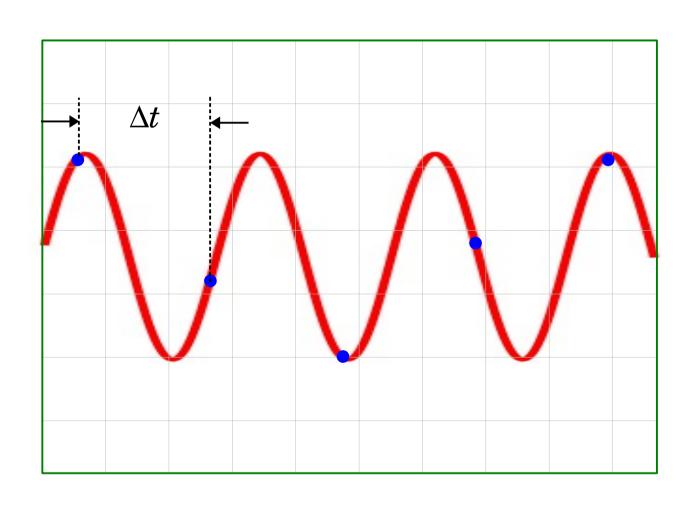
$$\Delta t \le \frac{1}{2\nu}$$



Nyquist theorem Sampling theorem

Temporal spacing of signal sampling

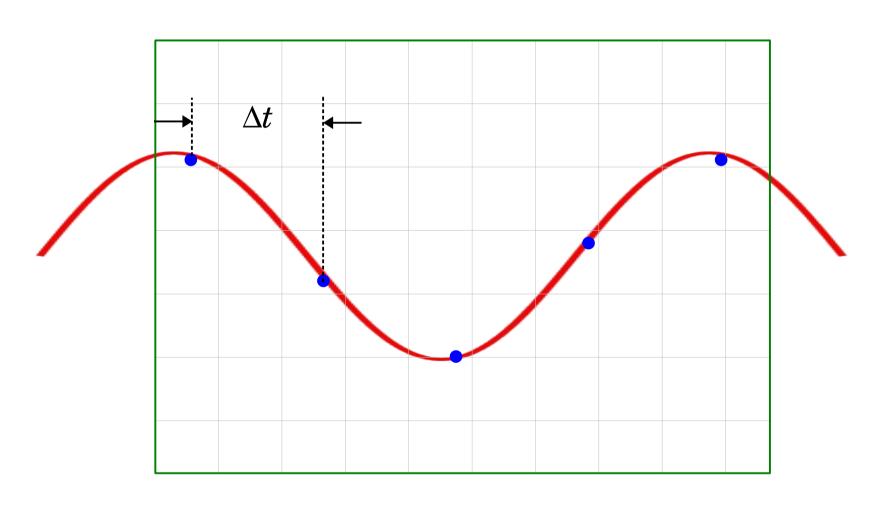
$$\Delta t > \frac{1}{2\nu}$$



Nyquist theorem Sampling theorem

Temporal spacing of signal sampling

$$\Delta t > \frac{1}{2\nu}$$



ALIASING

DIGITAL SCOPE: MEASUREMENT MENU

Period

Rise time

Frequency

• Fall time

• Average amplitude

• Duty cycle

Peak amplitude

RMS

Peak-to-peak amplitude

Max/Min signals

Horizontal and vertical adjustable cursors

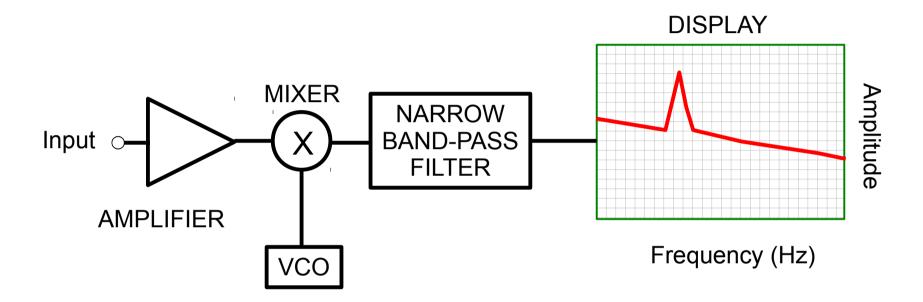
DIGITAL SCOPE: MATH MENU

Channel addition

Channel subtraction

Fast Fourier Transform (FFT):
Observe frequency spectrum of time signal

Spectrum Analyzer (Agilent N9320)



Operates like a radio with a very fast tuner

Spectrum Analyzer (Agilent N9320)

MENU SETUP

Start frequency Vertical scale (dB or linear)

Stop frequency Autoscale vertical axis

Averaging Preamp available

Markers Peak search

Operating range: 9 kHz – 3 GHz

Auto-tune rarely works

You should estimate where the expected signal will be